

What if there's a capacitor? Can we use the same tools?


$$\text{---} \text{---} \text{---} \quad V = \frac{Q}{C} \Rightarrow \frac{dV}{dt} = \frac{I}{C} \quad \text{or that } I = C \frac{dV}{dt}$$

So if the driver is $V(t) = \tilde{V} e^{i\omega t}$ then,

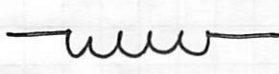
$$I = C i\omega \tilde{V} e^{i\omega t} = i\omega C V$$

This looks like $V = I "R"$ but this time the impedance is $Z_c = \frac{1}{i\omega C} = \frac{-i}{\omega C}$ (we use Z to indicate complex impedance)

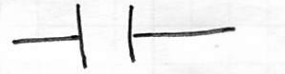
Summary

$$Z = R$$


R

$$Z = i\omega L$$


L

$$Z = -i/\omega C$$


C

General Result! you can treat each passive element like a simple resistor with the impedances given above. You can construct Z_{eff} using the standard "rules" for resistors.

In series,

$$Z_{\text{eff}} = Z_1 + Z_2 + Z_3 + \dots$$

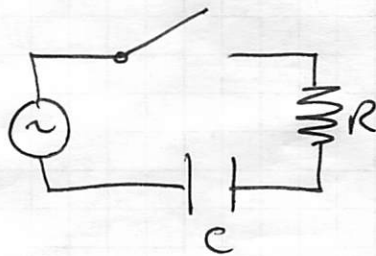
In parallel,

$$\frac{1}{Z_{\text{eff}}} = \frac{1}{Z_1} + \frac{1}{Z_2} + \frac{1}{Z_3} + \dots$$

Can apply usual Kirchhoff's Laws to them as well.

Example! An RC circuit

Turn it on @ $t = 0$
with $V(t) = V_0 \cos \omega t$



think of this as two Z 's,

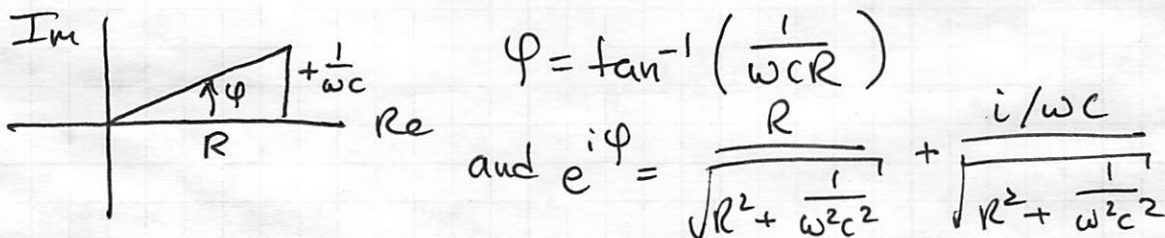
$$Z_{\text{tot}} = Z_R + Z_C = R - i/\omega C$$

with, $\tilde{V} = \tilde{I} Z \Rightarrow \tilde{I} = \frac{\tilde{V}}{R - i/\omega C}$ and $I_{\text{real}} = V_0 \operatorname{Re} \left(\frac{e^{i\omega t}}{R - i/\omega C} \right)$

the first part uses this method again,

$$\tilde{I} = V_0 \operatorname{Re} \left(\frac{e^{i\omega t}}{R - i/\omega C} \frac{R + i/\omega C}{R + i/\omega C} \right) = \frac{V_0}{R^2 + \frac{1}{\omega^2 C^2}} \operatorname{Re} \left(e^{i\omega t} (R + \frac{i}{\omega C}) \right)$$

Use the second method, draw a picture,



So the particular solution is,

$$I_p = \frac{V_0}{\sqrt{R^2 + \frac{1}{\omega^2 C^2}}} \cos(\omega t + \varphi)$$

We still need to solve the homogeneous equation,

$$0 = IR + Q/C \quad \Rightarrow \quad \frac{dI}{dt} R = -\frac{I}{C}$$

take time derivative
($dQ/dt = I$)

So, $\frac{dI}{I} = -\frac{1}{RC} dt \Rightarrow I_H = I_0 e^{-t/RC}$

So

$$I_{\text{tot}} = \underbrace{\frac{V_0}{\sqrt{R^2 + \frac{1}{\omega^2 C^2}} \cos(\omega t + \varphi)}_{\text{Steady solution}} + \underbrace{I_0 e^{-t/RC}}_{\substack{\text{transient} \\ \text{dies off} \\ \text{TPBD} \\ \text{by initial} \\ \text{conditions.}}}$$

Initial Conditions: ΔV_{cap} cannot suddenly change.

If it was zero before then just after $V_0 = IR$ b/c $\Delta V_{cap} = 0$ for just a quick moment. So $I(t=0) = V_0/R$.

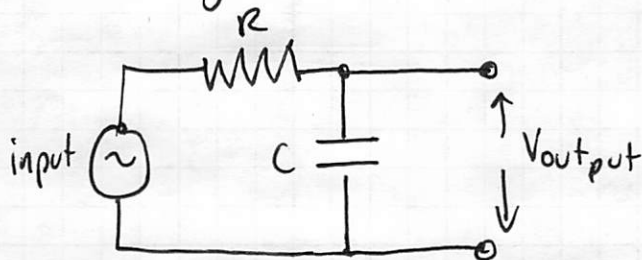
So, this gives,

$$I(t) = \frac{V_0}{\sqrt{R^2 + 1/\omega^2 C^2}} \cos(\omega t + \varphi) + \left(\frac{V_0}{R} - \frac{V_0 \cos \varphi}{\sqrt{R^2 + 1/\omega^2 C^2}} \right) e^{-t/RC}$$

Picked so @ $t=0$ $I = V_0/R$.

- time constant is "RC"
- After waiting several RC times the circuit oscillates with driver frequency, ω
- If $\omega \rightarrow 0$, $I_{long\ term} \rightarrow 0$, the capacitor blocks steady current.
- if $\omega \rightarrow \infty$, $I_{long\ term} \rightarrow \frac{V_0}{R} \cos(\omega t)$ like the capacitor isn't there.

It's often useful to put this setup in a circuit where you can read out a voltage (as a signal)



We know I in this circuit and V_{out} is,
 $V_{out} = I Z_C = I (-i/\omega C)$

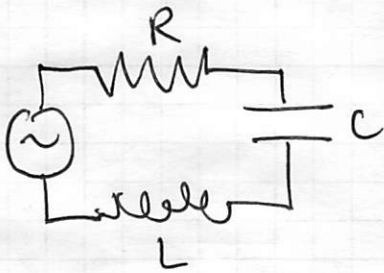
$$V_{out} = \frac{V_{in}}{Z_{tot}} Z_C = \frac{V_{in}}{R - \frac{i}{\omega C}} (-i/\omega C) = V_{in} \left(\frac{1}{1 + i\omega RC} \right)$$

with $\omega \rightarrow 0$, $V_{out} = V_{in}$, cap does nothing

with $\omega \rightarrow \infty$, $V_{out} \rightarrow 0$, "low pass filter"

Allows low frequencies to pass; suppresses high frequency.

with the phasor method & impedance, any circuit is basically a 184 circuit.



just use $Z = R + i\omega L - \frac{i}{\omega C}$

and then $\tilde{V} = \tilde{I} \tilde{Z}$